

ECE 2C Mid-Term Exam

May 9, 2013

Do not open exam until instructed to.

Closed book: 2 pages personal notes permitted

There are xx problems on this exam, and you have 75 minutes.

Use any and all reasonable approximations (5% accuracy is fine.), **AFTER STATING and approximately Justifying them.**

(1) *Answers without units are wrong.*

(2) *Plots without correctly labelled and dimensioned axes are wrong.*

Name: solution - even "a"

Perm #: _____

Problem	Points Received	Points Possible
1a		5
1b		5
1c		10
2a		5
2b		5
3a		5
3b		10
3c		10
4a		10
4b		10
5a	///	10
5b	///	10
total		100 75

Problem 1, 15 points

Part a, 5 points.

Q1 is a mobility-limited FET, ie.

$$I_d = (\mu C_{ox} W_g / 2L_g)(V_{gs} - V_{th})^2(1 + \lambda V_{ds})$$

where $(\mu C_{ox} W_g / 2L_g) = 1 \text{ mA/V}^2$, $\lambda = 0.0 \text{ V}^{-1}$, and $V_{th} = 0.3 \text{ V}$.

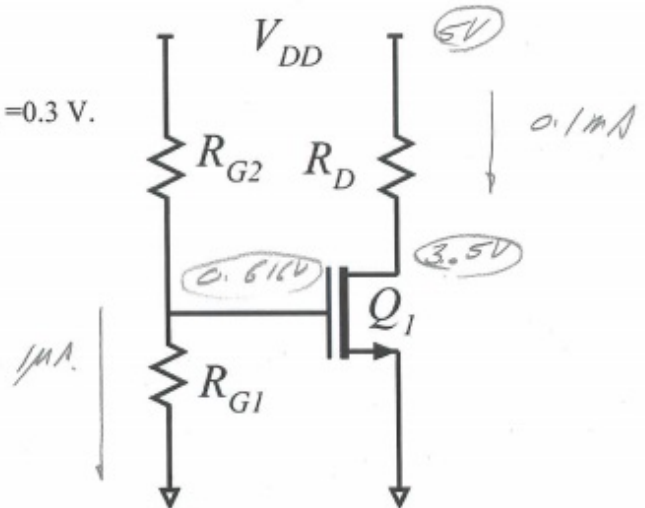
$V_{DD} = 5 \text{ Volts}$

The drain is biased at +3.5 Volts.

The drain current is 0.1 mA.

The DC current in R_{G1} is $1 \mu\text{A}$

Find R_{G1} , R_{G2} , R_D , and the DC gate voltage.



$$R_{G1} = \underline{616\Omega} \quad R_{G2} = \underline{4.58\text{M}\Omega} \quad R_D = \underline{15\text{k}\Omega}$$

$$V_{G1} = \underline{0.616\text{V}}$$

1 pt $\left[R_D = (5\text{V} - 3.5\text{V}) / 0.1\text{mA} = 1.5\text{V} / 0.1\text{mA} = 15\text{k}\Omega \right]$

2 pt $\left[\begin{aligned} 0.1\text{mA} &= \frac{1\text{mA}}{\text{V}^2} (V_{gs} - 0.3\text{V})^2 \rightarrow 0.1\text{V}^2 = (V_{gs} - 0.3\text{V})^2 \\ \rightarrow V_{gs} &= 0.616\text{V} \end{aligned} \right]$

1 pt $\left[R_{G1} = \frac{0.616\text{V}}{1\mu\text{A}} = 616\Omega \right]$

1 pt $\left[R_{G2} = \frac{5\text{V} - 0.616\text{V}}{1\mu\text{A}} = \frac{4.584\text{V}}{1\mu\text{A}} = 4.58\text{M}\Omega \right]$

Part b, 5 points

Using the FET parameters and DC bias conditions of part (a), draw a small-signal model of the FET and give numerical values of all small-signal parameters.

FET operating at $V_{gs} = 0.616V$, $V_{ds} = 3.5$

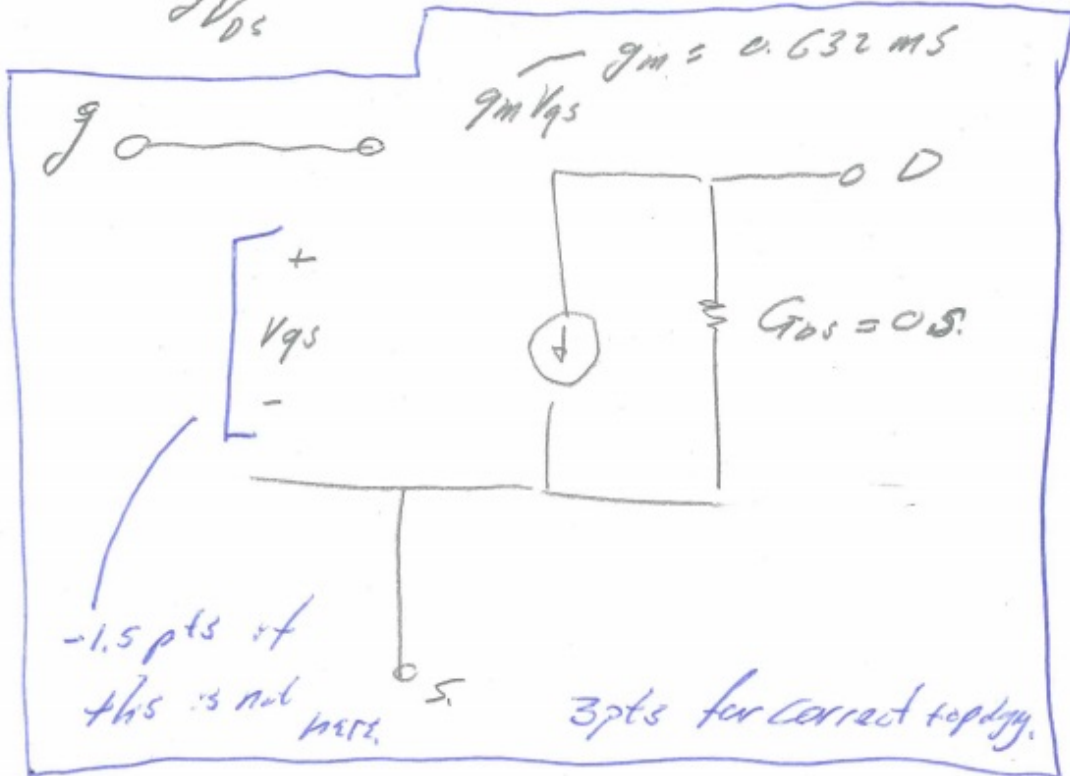
$$I_{d1} = \frac{1mA}{\sqrt{2}} \cdot (V_{gs} - V_{t1})^2 (1 + \lambda V_{ds})$$

$\begin{matrix} 0.3V \\ | \\ 1 \\ | \\ 0 \end{matrix}$

$$g_m \triangleq \frac{\partial I_D}{\partial V_{gs}} = \frac{2mA}{\sqrt{2}} \cdot (V_{gs} - 0.3V) = \underline{0.632 mS}$$

$\begin{matrix} 0.616 \\ | \\ 0.3V \end{matrix}$

$$g_{ds} = \frac{\partial I_D}{\partial V_{ds}} = 0 \text{ because } \lambda = 0$$

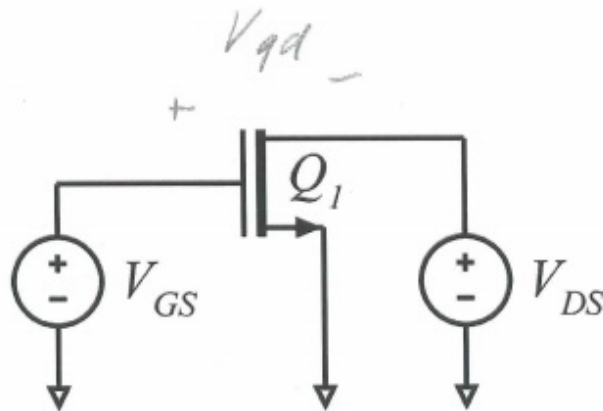


Part c, 10 points.

When FET gate lengths become very short, the neither the mobility-limited nor the velocity-limited models are correct. Instead, we have *degenerate/ballistic* characteristics, which for *low V_{ds}* gives:

$$I_d = K(V_{gs} - V_{th})^{3/2} - K(V_{gd} - V_{th})^{3/2}$$

For this FET, $K=1 \text{ mA/V}^{3/2}$, and $V_{th}=0.2 \text{ V}$.



In the circuit shown, the gate is biased at +0.8Volts, and the drain is biased at +0.2 Volts. **Draw a small-signal model of the FET** labelling all elements and all control voltages, **and give numerical values of all small-signal parameters.**

$$I_d = K(V_{gs} - V_{th})^{3/2} - K(V_{gd} - V_{th})^{3/2}$$

key: $V_{gd} = V_{gs} - V_{ds}$ (important!)

$$I_d = K(V_{gs} - V_{th})^{3/2} - K(V_{gs} - V_{ds} - V_{th})^{3/2}$$

$$g_m = \frac{\partial I_d}{\partial V_{gs}} = \frac{3K}{2}(V_{gs} - V_{th})^{1/2} - \frac{3K}{2}(V_{gs} - V_{ds} - V_{th})^{1/2}$$

$$= \frac{3 \text{ mA}}{2 \text{ V}^{1/2}} (0.8\text{V} - 0.2\text{V})^{1/2} - \frac{3 \text{ mA}}{2 \text{ V}^{1/2}} (0.8\text{V} - 0.4\text{V})^{1/2} = 0.213 \text{ mS}$$

$$g_{ds} = \frac{\partial I_D}{\partial V_{DS}} = (+) \frac{3K}{2}(V_{gs} - V_{ds} - V_{th})^{1/2} = \frac{3 \text{ mA}}{2 \text{ V}^{1/2}} (0.8\text{V} - 0.4\text{V})^{1/2} = 0.95 \text{ mS}$$



①

Problem 2, 15 points

Q1 is a mobility-limited FET, ie.

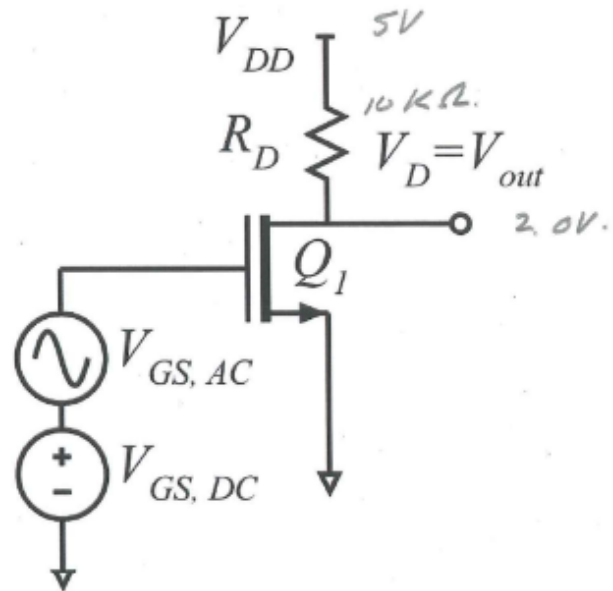
$$I_d = (\mu C_{ox} W_g / 2L_g) (V_{gs} - V_{th})^2 (1 + \lambda V_{ds})$$

where $(\mu C_{ox} W_g / 2L_g) = 1 \text{ mA/V}^2$,

$\lambda = 0.1 \text{ V}^{-1}$, and $V_{th} = 0.3 \text{ V}$.

$V_{DD} = 5 \text{ Volts}$

$R_D = 10 \text{ k}\Omega$



Part a, 5 points

We wish to have a DC drain voltage of 2.0 Volts.

What DC gate-source voltage does this require?

$V_{GS,DC} = \underline{0.8 \text{ V}}$

2
$$I_D = \frac{5\text{V} - 2\text{V}}{10\text{k}\Omega} = \frac{3\text{V}}{10\text{k}\Omega} = 0.3 \text{ mA}$$

3
$$I_{d1} = 0.3 \text{ mA} = \frac{1 \text{ mA}}{\text{V}^2} (V_{gs} - 0.3\text{V})^2 \left(1 + \frac{V_{ds}}{10\text{V}}\right)$$

$$0.25 \text{ V}^2 = (V_{gs} - 0.3\text{V})^2 \rightarrow V_{gs} = 0.8 \text{ V}$$

Part b, 5 points

$$V_{GS,AC}(t) = 1 \text{ mV} \cdot \cos(2\pi \cdot 1\text{kHz} \cdot t)$$

What is the AC drain voltage?

$$V_{D,AC}(t) = \underline{\hspace{2cm}}$$

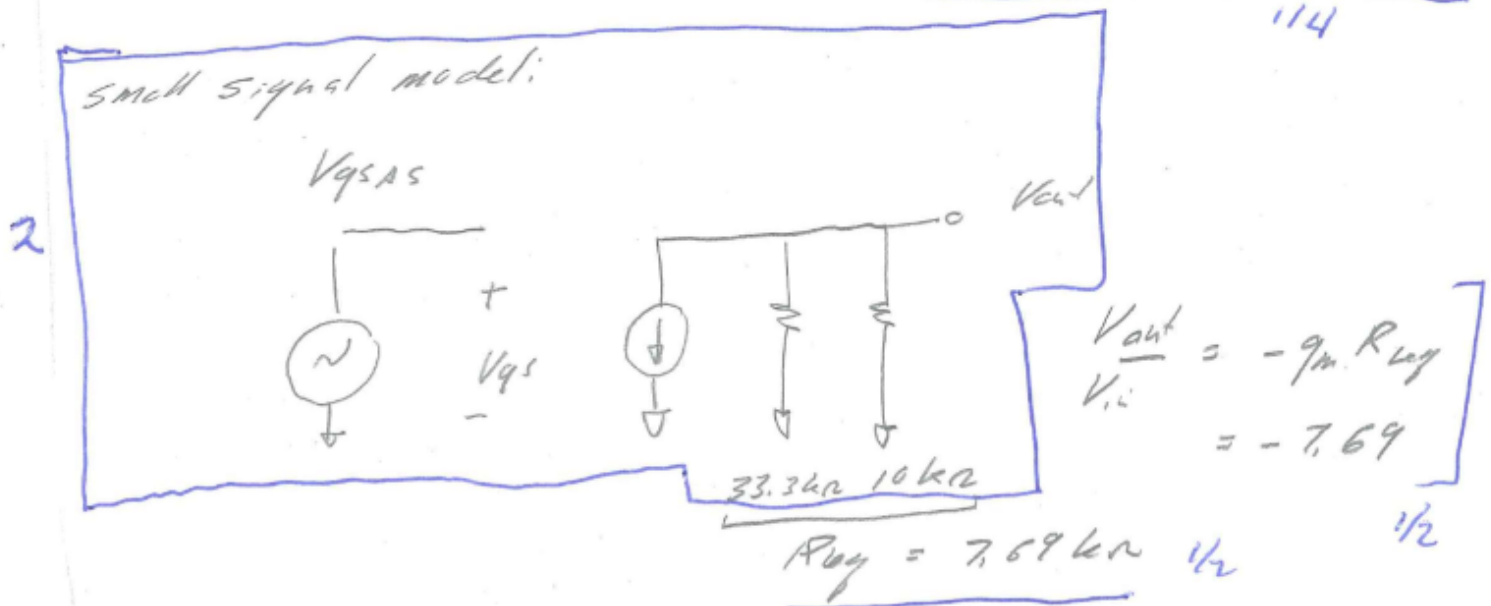
we need ss model (!)

ok to drop

$$1/4 \left[g_m \approx 2 \mu C_{ox} Wg / (2Lg) (V_{GS} - V_{th}) (1 + \lambda V_{DS}) \right]$$

$$1/4 \left[= \frac{2 \text{ mA}}{V^2} (0.8\text{V} - 0.3\text{V}) = \underline{1 \text{ mS}} \right]$$

$$1/4 \left[r_{ds} \approx \frac{\partial I_D}{\partial V_{DS}} = \frac{I_D}{1 + \lambda V_{DS}} \approx \lambda I_D = \frac{0.3 \text{ mA}}{10\text{V}} = 30 \mu\text{S} = 33.3 \text{ k}\Omega \right]$$

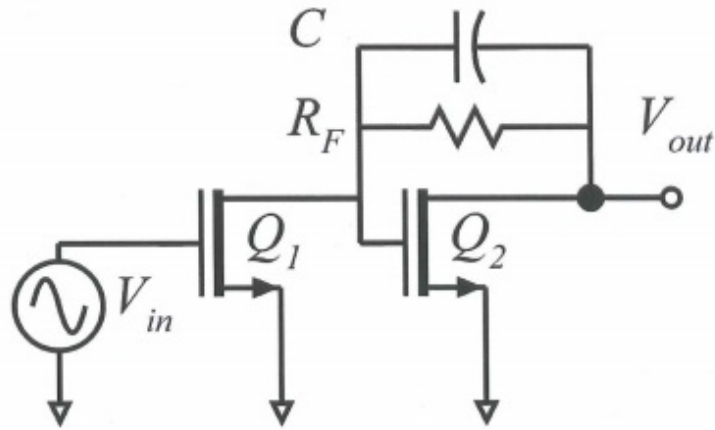


$$1 \left[V_{out}(t) = -7.69 \text{ mV} \cdot \cos(2\pi \cdot 1\text{kHz} \cdot t) \right]$$

Problem 3, 25 points

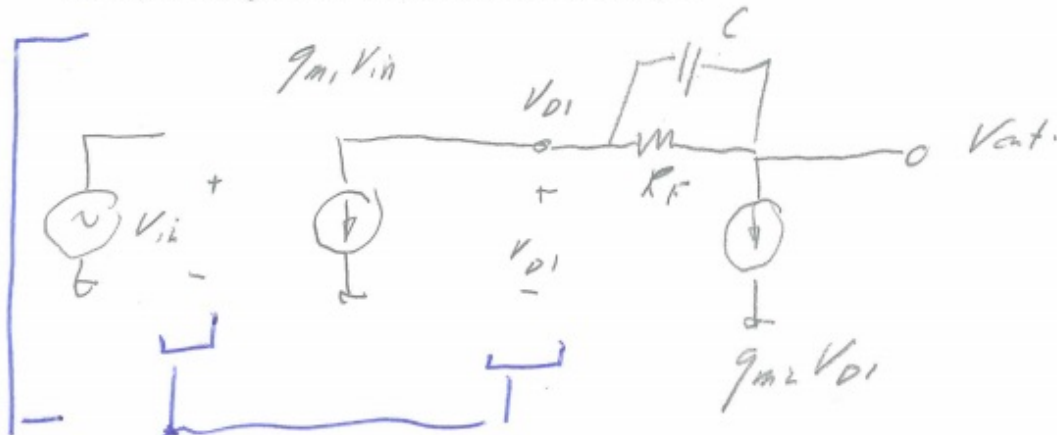
You will be working with the circuit at right. Ignore DC bias. You don't need it.

The transistors Q1 and Q2 have transconductances g_{m1} and g_{m2} , both nonzero, but $G_{ds1} = G_{ds2} = 0$ mS. The transistors have no gate-source or gate-drain capacitances.



Part a, 5 points

replacing the transistors with their small-signal models, draw a small-signal equivalent circuit, labelling all elements and all control voltages.



topology must be correct
 component labels must be present
 these must be labelled.

Part b, 10 points

Compute $V_{out}(s)/V_{in}(s)$: The answer must be in one of these two standard forms:

$$\frac{V_{out}(s)}{V_{in}(s)} = \frac{V_{out}}{V_{in}} \Big|_{\text{low-frequency value}} \times \frac{1 + b_1s + b_2s^2 + \dots}{1 + a_1s + a_2s^2 + \dots} \quad \text{or}$$

$$\frac{V_{out}(s)}{V_{in}(s)} = \frac{V_{out}}{V_{in}} \Big|_{\text{low-frequency value}} \times \frac{(1 - s/s_{z1})(1 - s/s_{z2})\dots}{(1 - s/s_{p1})(1 - s/s_{p2})\dots}$$

$$V_{out}(s)/V_{in}(s) = \underline{\hspace{10em}}$$

3 $\Sigma I = 0 @ V_{D1}$ $\rightarrow g_{m1} V_{in} + V_{D1}(\Delta C + G_F) + V_{out}(-\Delta C - G_F) = 0$
 $\rightarrow V_{D1}(\Delta C + G_F) + V_{out}(-\Delta C - G_F) = -g_{m1} V_{in}$

3 $\Sigma I = 0 @ V_{out}$ $\rightarrow V_{out}(\Delta C + G_F) + V_{D1}(g_{m2} - \Delta C - G_F) = 0$

$$\begin{bmatrix} \Delta C + G_F & -\Delta C - G_F \\ g_{m2} - \Delta C - G_F & \Delta C + G_F \end{bmatrix} \begin{bmatrix} V_{D1} \\ V_{out} \end{bmatrix} = \begin{bmatrix} -g_{m1} V_{in} \\ 0 \end{bmatrix}$$

$$D(s) = \begin{vmatrix} sC + G_F & -sC - G_F \\ g_{m2} - sC - G_F & sC + G_F \end{vmatrix} = g_{m2}(sC + G_F)$$

$$N(s) = \begin{vmatrix} v_i & -g_{m1} v_i \\ g_{m2} - sC - G_F & 0 \end{vmatrix} = g_{m1}(g_{m2} - sC - G_F)$$

$$= \frac{g_{m1}(g_{m2} - sC - G_F)}{g_{m2}(sC + G_F)}$$

$$= \frac{g_{m1}(g_{m2} - G_F)}{g_{m2} G_F} \frac{(1 - sC/(g_{m2} - G_F))}{1 + sC/G_F}$$

$$\frac{V_o(s)}{V_{in}(s)} = \frac{g_{m1}(g_{m2} R_F - 1)}{g_{m2}} \frac{1 - sC(g_{m2} - G_F)^{-1}}{1 + sC R_F}$$

②

Part c, 10 points

Now $g_{m1} = g_{m2} = 1 \text{ ms}$, $R_f = 10 \text{ k}\Omega$, $C = 1 \text{ nF}$.

Give the frequencies, in Hz, of any and all poles and zeros. Clearly indicate if any lie in the right half of the S-plane

1st pole frequency = 15.9 kHz, 2nd pole frequency = _____,
3rd pole frequency = _____

1st zero frequency = 1436 Hz, 2nd zero frequency = _____,
3rd zero frequency = _____

poles:

$$3 \left[\tau = C \cdot R_f = 1 \text{ nF} \cdot 10 \text{ k}\Omega = 10 \mu\text{s} \right]$$

$$2 \left[\tau_{TP} = \frac{0.159}{10 \mu\text{s}} = 15.9 \text{ kHz} \right]$$

zeros - in RHP.

$$3 \left[\tau = \frac{C}{g_{m2} - G_H} = \frac{1 \text{ nF}}{1 \text{ ms} - 0.1 \text{ ms}} = 1.111 \mu\text{s} \right]$$

$$2 \left[\tau_{zero} = \frac{0.159}{1.111 \mu\text{s}} = 1436 \text{ Hz} \right]$$

Problem 4, 20 points

We have a circuit for which

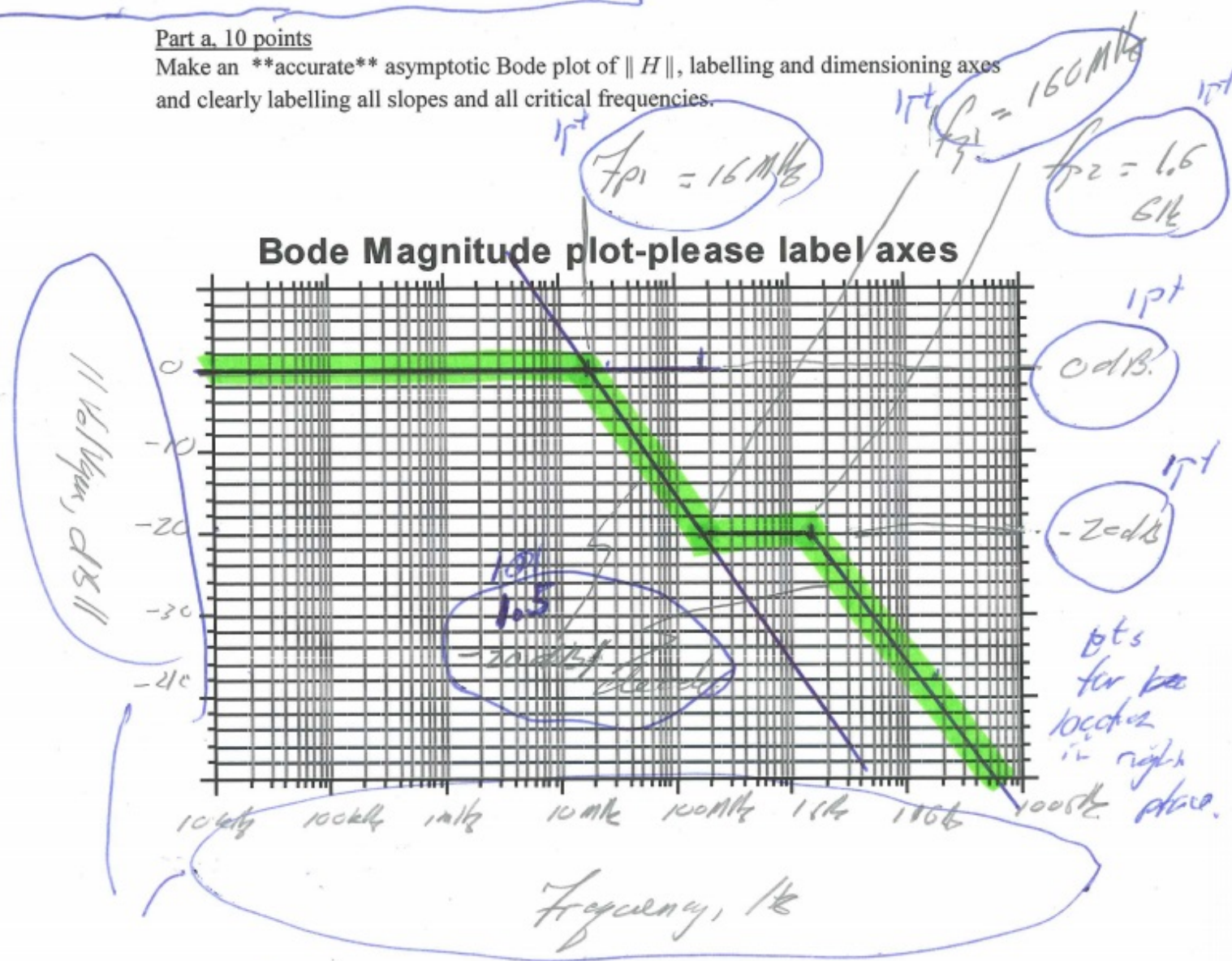
$$\frac{V_{out}(s)}{V_{in}(s)} = H(s) = H_{DC} \times \frac{1 + s\tau_1}{(1 + s\tau_2)(1 + s\tau_3)}$$

where $\tau_1 = 1 \text{ ns}$, $\tau_2 = 10 \text{ ns}$, $\tau_3 = 0.1 \text{ ns}$.

Handwritten notes in a box:
 zero @ 160 MHz
 poles: 16 MHz, 1.6 GHz
 1/2 pt each

Part a. 10 points

Make an **accurate** asymptotic Bode plot of $\|H\|$, labelling and dimensioning axes and clearly labelling all slopes and all critical frequencies.



correct scales & labels. - 2 pts.

a exam.

Part b, 10 points

Make an ****accurate**** root locus plot of $H(s)$, labelling and dimensioning axes and clearly labelling all critical frequencies.

be sure to label axes with #s and units

